

High Speed, Fast Settling Precision Operational Amplifier

OP-42

FEATURES

Fast

•	Siew Hate	50V/μs Min
•	Settling-Time (0.01%)	1us Max
•	Gain-Bandwidth Product	10MHz Typ
_		

Precise

•	100.30	
•	Common-Mode Rejection	88dB Min
•	Open-Loop Gain	500V/mV Min
•	Offset Voltage	750uV Max
	Bias Current	200pA May

Excellent Radiation Hardness Available in Die Form

ORDERING INFORMATION †

T, = 25°0	c		PACK	AGE	,	OPERATING
V _{OS} MAX (mV)		CERDIP 8-PIN	PLASTIC 8-PIN	SO 8-PIN	LCC 20-CONTACT	TEMPERATURE RANGE
1.0	OP42AJ*	OP42AZ*	_	_	OP42ARC/883	MIL
0.75	OP42EJ	OP42EZ		_	_	IND
1.5	OP42FJ	OP42FZ		_	_	IND
5.0		-	OP42GP	OP42GS	-	XIND

- For devices processed in total compliance to MIL-STD-883, add /883 after part number. Consult factory for 883 data sheet.
- † Burn-in is available on commercial and industrial temperature range parts in cerdip, plastic dip, and TO-can packages.

GENERAL DESCRIPTION

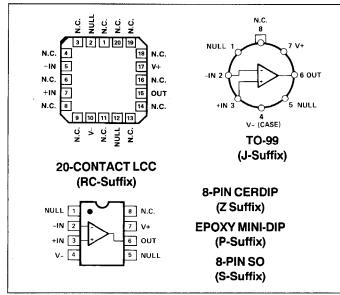
The OP-42 is a fast precision JFET-input operational amplifier. Similar in speed to the OP-17, the OP-42 offers a symmetric

 $58V/\mu s$ slew rate and is internally compensated for unity-gain operation. OP-42 speed is achieved with a supply current of less than 6mA. Unity-gain stability, a wide full-power bandwidth of 900kHz, and a fast settling-time of 800ns to 0.01% make the OP-42 an ideal output amplifier for fast digital-to-analog converters.

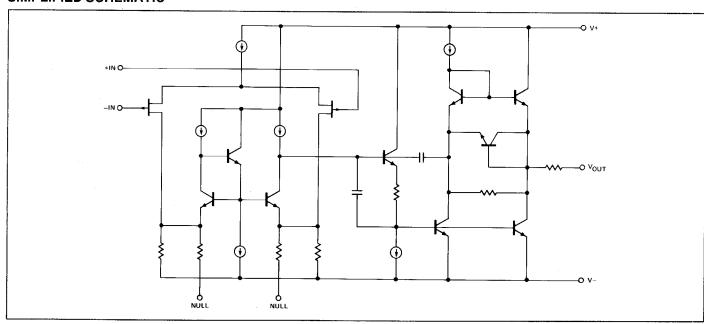
Equal attention was given to both speed and precision in the OP-42 design. Its tight $750\mu V$ maximum input offset voltage combined with well-controlled drift of less than $10\mu V/^{\circ}C$ eliminates the need for external nulling in many circuits. The OP-42's

Continued

PIN CONNECTIONS



SIMPLIFIED SCHEMATIC



GENERAL DESCRIPTION Continued

common-mode rejection of 88dB minimum over a $\pm 11V$ input voltage range is exceptional for a high-speed amplifier. High CMR combined with a minimum 500V/mV gain into $10k\Omega$ load ensure excellent linearity in both noninverting and inverting gain configurations. The low input bias and offset currents provided by the JFET input stage suit the OP-42 for use in high-speed sample and hold circuits, peak detectors, and log amplifiers. Excellent radiation hardness characteristics make the OP-42 ideal for military and aerospace applications.

The OP-42 conforms to the standard 741 pinout with nulling to V—. The OP-42 upgrades the performance of circuits using the AD544, AD611, AD711, and LF400 by direct replacement. In circuits without nulling, the OP-42 offers an upgrade for designs using the OP-16, OP-17, LT1022, LT1056, and HA2510.

ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage	±20V
Input Voltage (Note 2)	±20V
Differential Input Voltage (Note 2)	
Output Short-Circuit Duration	

Storage Temperature Range	65°C to +175°C
Operating Temperature Range	
OP42A (J, Z)	–55°C to +125°C
OP42E, F (J, Z)	25°C to +85°C
OP42G	40°C to +85°C
Junction Temperature	65°C to +175°C
Lead Temperature Range (Soldering,	

PACKAGE TYPE	Θ _{jA} (NOTE 3)	Θ _{IC}	UNITS
TO-99 (J)	150	18	°C/W
8-Pin Hermetic DIP (Z)	148	16	°C/W
8-Pin Plastic DIP (P)	103	43	°C/W
20-Contact LCC (RC,TC)	98	38	°C/W
8-Pin SO (S)	158	43	°C/W

NOTES:

- Absolute maximum ratings apply to both DICE and packaged parts, unless otherwise noted.
- 2. For supply voltages less than $\pm 20V$, the absolute maximum input voltage is equal to the supply voltage.
- Θ_{JA} is specified for worst case mounting conditions, i.e., Θ_{JA} is specified for device in socket for TO, CERDIP, P-DIP, and LCC packages; Θ_{JA} is specified for device soldered to printed circuit board for SO package.

ELECTRICAL CHARACTERISTICS at $V_S = \pm 15V$, $T_A = +25$ °C, unless otherwise noted.

			OP-42E			OP-42F			OP-42G			
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
Offset Voltage	v _{os}		-	0.3	0.75	_	0.4	1.5	_	1.5	5.0	mV
Input Bias Current	I _B	V _{CM} = 0V T _j = 25°C		80	200		130	250	-	130	250	pA
Input Offset Current	los	V _{CM} = 0V T _j = 25°C	-	4	40	_	6	50	_	6	50	pA
Input Voltage Range	IVR	(Note 1)	±11	+12.5 12.0	_	±11	+12.5 -12.0	_	±11	+12.5 -12.0	_	٧
Common-Mode Rejection	CMR	V _{CM} = ±11V	88	98	-	80	92	_	80	92	_	dB
Power-Supply Rejection Ratio	PSRR	V _S = ±10V to ±20V	-	9	40	-	12	50	-	12	50	μV/V
Large-Signal Voltage Gain	A _{VO}	$\begin{aligned} R_L &= 10k\Omega \\ R_L &= 2k\Omega \\ R_L &= 1k\Omega \end{aligned} \begin{aligned} V_0 &= \pm 10V \\ T_j &= 25^{\circ}C \end{aligned}$	500 200 100	900 260 170	- - -	500 200 100	900 260 170	- - -	500 200 100	900 260 170	-	V/mV
Output Voltage Swing	v _o	R _L = 1kΩ	±11.5	+12.5 -11.9	_	±11.5	+12.5 11.9	_	±11.5	+12.5 -11.9	_	V
Short-Circuit Current Limit	I _{sc}	Output Shorted to Ground	±20	+33 –28	±60	±20	+33 –28	±60	±20	+33 –28	±60	mA
Supply Current	Isy	No Load V _O = 0V	-	5.1	6.0	-	5.1	6.5	-	5.1	6.5	mA
Slew Rate	SR	- LANAVATO	50	58	_	40	50	_	40	50	_	V/μs
Full-Power Bandwidth	вw _р	(Note 2)	750	900	_	600	800	_	600	800	_	kHz
Gain-Bandwidth Product	GBW	f _O = 10kHz	_	10	_	_	10		_	10	-	MHz
Settling -Time	t _s	10V Step 0.01% (Note 3)	-	0.8	1.0	_	0.9	1.2	_	0.9	1.2	μs
Overload Recovery Time	ton		-	700	_		700		_	700	_	ns
Phase Margin	φο	0db Gain	_	47	-	_	47	_		47	_	degrees
Gain Margin	A ₁₈₀	180° Open-Loop Phase Shift	-	9	_		9	_	_	9	_	dB
Capacitive Load Drive Capability	CL	Unity-Gain Stable (Note 4)	100	300	_	100	300	_	100	300	_	pF

ELECTRICAL CHARACTERISTICS at $V_S = \pm 15V$, $T_A = +25$ °C, unless otherwise noted. *Continued*

				OP-42E			OP-42	=	OP-42G			
PARAMETER	SYMBOL	SYMBOL CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
Differential Input Impedance	Z _{IN}		-	10 ¹² 6	-	_	10 ¹² 6	_	_	10 ¹² 6	-	Ω[pF
Open-Loop Output Resistance	R _O		_	50	_	_	50	-	_	50	_	Ω
Voltage Noise	e _{n p-p}	0.1Hz to 10Hz	-	2	-		2	_		2	_	μV _{p-p}
		f _O = 10Hz	_	38		_	38	_	_	38	_	
Voltage Noise	e _n	f _O = 100Hz	-	16	_	_	16	-	_	16	-	nV/√Hz
Density		f _O = 1kHz		13		_	13	_	_	13	-	HV/√ F1Z
		f ₀ = 10kHz	-	12	-		12	-	_	12	-	
Current Noise Density	in	f _O = 1kHz	_	0.007		_	0.007	_	_	0.007		pA/√Hz
External V _{OS} Trim Range		R _{pot} = 20kΩ	_	4	_	-	4	_	_	4	_	m∨
Long-Term V _{OS} Drift			_	5	_	_	5	_	-	5	_	μV/month
Supply Voltage Range	v _s		±8	±15	±20	±8	±15	±20	±8	±15	±20	٧

NOTES:

ELECTRICAL CHARACTERISTICS at $V_S = \pm 15V$, $T_A = +25$ °C, unless otherwise noted.

				OP-42A		
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Offset Voltage	Vos			0.3	1.0	mV
Input Bias Current	l _e	V _{CM} = 0V T _j = 25°C	_	80	200	pΑ
Input Offset Current	los	V _{CM} = 0V T _j = 25°C	_	4	40	pA
Input Voltage Range	IVR	(Note 1)	±11	+12.5 12.0	_	V
Common-Mode Rejection	CMR	V _{CM} = ±11V	86	96	_	dB
Power-Supply Rejection Ratio	PSRR	V _S = ±10V to ±20V	_	9	40	μV/V
Large-Signal		F ₁ = 10kΩ V = -10V	500	900	_	
Voltage Gain	A_{VO}	$R_1 = 2k\Omega$ $V_0 = 10V$	200	260	_	V/mV
		$R_L = 1k\Omega$ $i = 25 C$	100	170	_	
Output Voltage Swing	v _o	$H_L = 1k\Omega$	±11.5	+12.5 -11.9	-	V
Short-Circuit Current Limit	l _{sc}	Output Shorted to Ground	±20	+33 28	±60	mA
Supply Current	Isy	No Load V _O = 0V	-	5.1	6.0	mA
Slew Rate	SR		45	52	_	V/μs
Full-Power Bandwidth	BW _p	(Note 2)	700	850	-	kHz
Gain-Bandwidth Product	GBW	f _O = 10kHz	_	10	_	MHz
Settling -Time	t _s	10V Step 0.01% (Note 3)	-	0.8	1.0	μs
Overload Recovery Time	t _{o n}		_	700	_	ns
Phase Margin	Фо	0db Gain	_	47		degrees

^{1.} Guaranteed by CMR test. 2. Guaranteed by slew-rate test and formula $BW_p = SR/(2\pi 10V_{PEAK})$.

Settling-time is sample tested for A and E grades. Test circuit is shown in Figure 4. Settling-time for F grade is guaranteed but not tested.
 Guaranteed but not tested.

OP-42 ELECTRICAL CHARACTERISTICS at $V_s = \pm 15V$, $T_A = 25^{\circ}C$, unless otherwise noted. *Continued*

				OP-42A		
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Gain Margin	A ₁₈₀	180° Open-Loop Phase Shift	-	9	_	dB
Capacitive Load Drive Capability	CĽ	Unity-Gain Stable (Note 4)	100	300	_	pF
Differential Input Impedance	Z _{IN}			10 ¹² 6	_	Ω pF
Open-Loop Output Resistance	R _O		-	50	-	Ω
Voltage Noise	e _{n p-p}	0.1Hz to 10Hz	. –	2	_	μV _{p-p}
Voltage Noise Density	e _n	f _O = 10Hz f _O = 100Hz f _O = 1kHz f _O = 10kHz	- - - -	38 16 13 12	- - -	nV/√Hz
Current Noise Density	i _n	f _O = 1kHz	_	0.007	_	pA/√Hz
External V _{OS} Trim Range		R _{pot} = 20kΩ	-	4	_	mV
Long-Term V _{OS} Drift			-	5	_	μV/month
Supply Voltage Range	v _s		±8	±15	±20	V

NOTES:

ELECTRICAL CHARACTERISTICS at $V_S = \pm 15V$, $-25^{\circ}C \le T_A \le 85^{\circ}C$ for E/F grades, and $-40^{\circ}C \le T_A \le +85^{\circ}C$ for G grade, unless otherwise noted.

			OP-42F OP-42F		OP-42G							
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
Offset Voltage	Vos		_	0.4	1.2	_	0.6	2.5	-	2.0	6.0	mV
Offset Voltage Temperature Coefficient	TCVos		-	4	10	_	8		-	8	-	μV/°C
Input Bias Current	I _B	(Note 1)	-	0.5	1.2	_	0.6	2.0	_	0.6	2.0	nA
Input Offset Current	los	(Note 1)	_	0.05	0.2		0.06	0.4	_	0.06	0.4	nA
Input Voltage Range	IVR	(Note 2)	±11	+12.5 -12.0	_	±11	+12.5 -12.0	_	±11	+12.5 -12.0	-	V
Common-Mode Rejection	CMR	V _{CM} = ±11V	86	96	_	80	94	-	80	94	_	dB
Power-Supply Rejection Ratio	PSRR	V _S = ±10V to ±20V	_	2	40	_	6	50	_	6	50	μ V/ V
Large-Signal Voltage Gain	A _{vo}	$R_L = 10k\Omega$ (Note 1) $R_L = 2k\Omega$ $V_O = \pm 10V$	200 100	500 160	_	200 100	500 160	-	200 100	500 160	_	V/mV
Output Voltage Swing	v _o	R _L = 2kΩ	±11.0	+12.3 -11.8	-	±11.0	+12.3 -11.8	_	±11.0	+12.3 -11.8		٧
Short-Circuit Current Limit	I _{sc}	Output Shorted to Ground	±8	-	±60	±8	_	±60	±8	-	±60	mA
Supply Current	I _{SY}	No Load V _O = 0V	_	5.1	6.0	_	5.1	6.5	_	5.1	6.5	mA
Slew Rate	SR	R _L = 2kΩ	45	57	-	40	50	-	40	50	-	V/µs
Capacitive Load Drive Capability	CL	Unity-Gain Stable (Note 3)	100	250		100	250	-	100	250	_	pF

NOTES

^{1.} Guaranteed by CMR test.

^{2.} Guaranteed by slew-rate test and formula ${\rm BW}_{\rm p} = {\rm SR}/(2\pi 10 {\rm V}_{\rm PEAK}).$

Settling-time is sample tested for A and E grades. Test circuit is shown in Figure 4. Settling-time for F grade is guaranteed but not tested.

^{4.} Guaranteed but not tested.

^{1.} T_i = 85°C for E/F/G Grades; T_i = 125°C for A grade.

^{2.} Guaranteed by CMR test.

^{3.} Guaranteed but not tested.

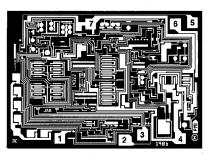
ELECTRICAL CHARACTERISTICS at $V_S = \pm 15V$, $-55^{\circ}C \le T_A \le 125^{\circ}C$ for A grade, unless otherwise noted.

				OP-42A		
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Offset Voltage	V _{os}		_	0.5	2.0	mV
Offset Voltage Temperature Coefficient	TCV _{os}		_	4	10	μV/°C
Input Bias Current	I _B	(Note 1)	-	6	20	nA
Input Offset Current	los	(Note 1)	-	0.2	1.0	nA
Input Voltage Range	IVR	(Note 2)	±11	+12.5 -12.0	_	٧
Common-Mode Rejection	CMR	V _{CM} = ±11V	80	94	_	dB
Power-Supply Rejection Ratio	PSRR	V _S = ±10V to ±20V	-	10	50	μV/V
Large-Signal Voltage Gain	A _{vo}	R _L = 10kΩ (Note 1) R _L = 2kΩ $V_0 = \pm 10V$	160 80	350 110		V/mV
Output Voltage Swing	v _o	R _L = 2kΩ	±11.0	+12.3 -11.8	_	٧
Short-Circuit Current Limit	^I sc	Output Shorted to Ground	±8	_	±60	mA
Supply Current	Isy	No Load V _O = 0V		5.1	6.0	mA
Slew Rate	SR	R _L = 2kΩ	40	52	-	V/µs
Capacitive Load Drive Capability	C _L	Unity-Gain Stable (Note 3)	100	250	_	pF

^{1.} T_j = 85°C for E/F Grades; T_j = 125°C for A grade. 2. Guaranteed by CMR test.

^{3.} Guaranteed but not tested.

DICE CHARACTERISTICS



DIE SIZE 0.098×0.070 inch, 6860 sq. mils (2.49 \times 1.78 mm, 4.43 sq. mm)

- 1. OFFSET VOLTAGE NULL
- 2. INVERTING INPUT
- 3. NONINVERTING INPUT
- 4. NEGATIVE SUPPLY
- 5. OFFSET VOLTAGE NULL
- 6. AMPLIFIER OUTPUT
- 7. POSITIVE SUPPLY

WAFER TEST LIMITS at $V_S = \pm 15 V$, $T_j = 25 ^{\circ} C$, unless otherwise noted.

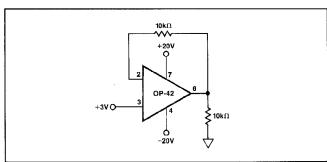
PARAMETER	SYMBOL	CONDITIONS	OP-42N LIMIT	UNITS
Input Bias Current	I _B	V _{CM} = 0V	250	pA MAX
Input Offset Current	Ios	V _{CM} = 0V	50	pA MAX
Input Voltage Range	IVR	(Note 1)	±11	V MIN
Common-Mode Rejection	CMR	$V_{CM} = \pm 11V$	80	dB MIN
Power-Supply Rejection Ratio	PSRR	$V_S = \pm 10V$ to $\pm 20V$	50	μV/V MAX
Large-Signal Voltage Gain	A _{vo}	$egin{aligned} R_L &= 10k\Omega \ R_L &= 2k\Omega \ R_L &= 1k\Omega \end{aligned}$	500 200 100	V/mV MIN
Output Voltage Swing	v _o	$R_L = 1k\Omega$	±11.5	V MIN
Short-Circuit Current Limit	l _{sc}	Output Shorted to Ground	±20/±60	mA MIN/MAX
Supply Current	I _{SY}	No Load V _O = 0V	6.5	mA MAX
Slew Rate	SR		40	V/μs MIN
Capacitive Load Drive Capability	CL	Unity-Gain Stable (Note 2)	100	pF MIN

NOTES:

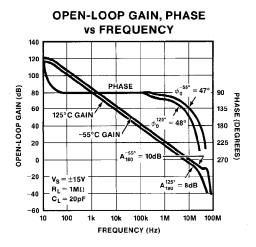
- 1. Guaranteed by CMR test.
- 2. Guaranteed but not tested.

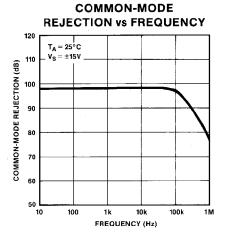
Electrical tests are performed at wafer probe to the limits shown. Due to variations in assembly methods and normal yield loss, yield after packaging is not guaranteed for standard product dicc. Consult factory to negotiate specifications based on dice lot qualification through sample lot assembly and testing.

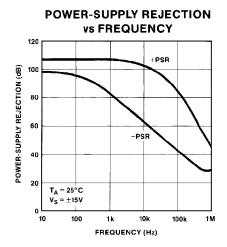
BURN-IN CIRCUIT

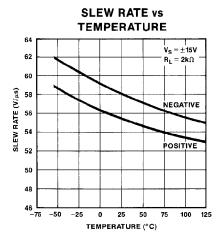


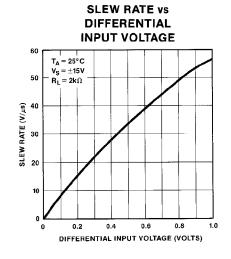
TYPICAL PERFORMANCE CHARACTERISTICS

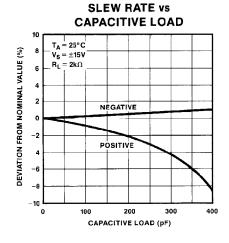


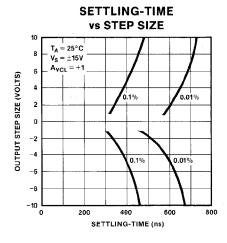


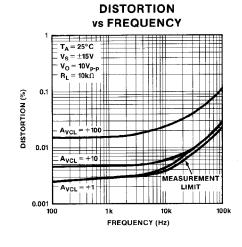


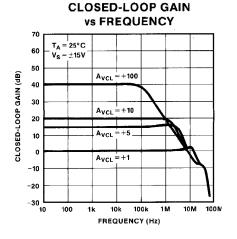




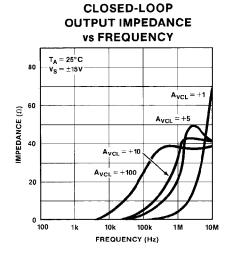


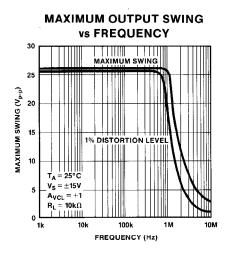


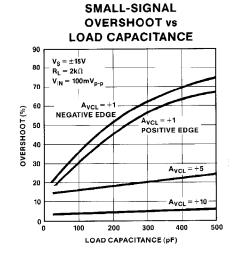


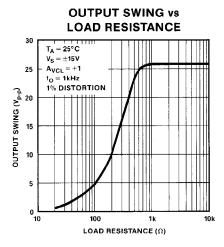


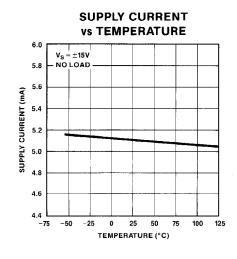
TYPICAL PERFORMANCE CHARACTERISTICS

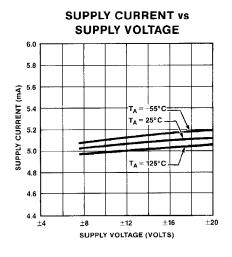


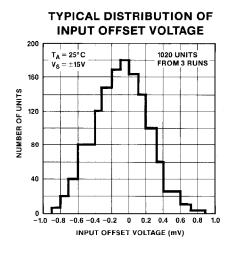


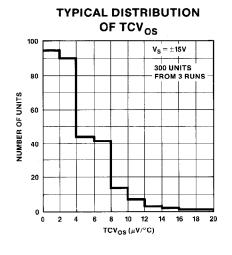


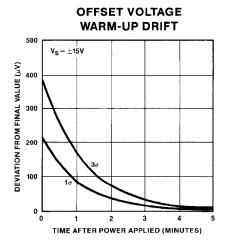




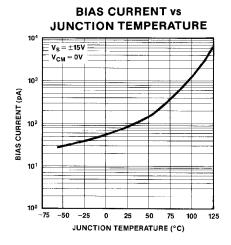


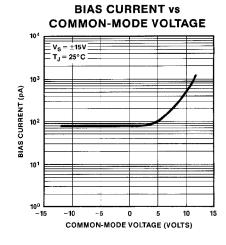


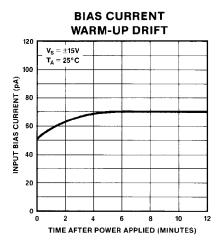


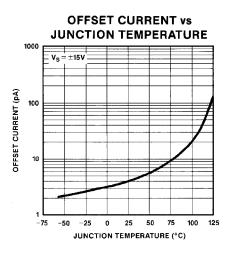


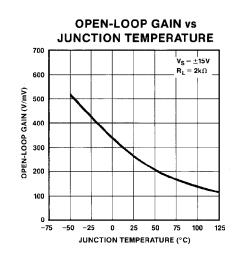
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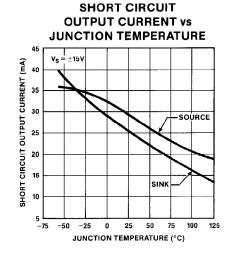


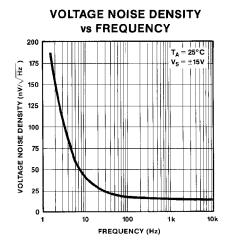












APPLICATIONS INFORMATION

The OP-42 combines speed with a high level of input precision usually found only with slower devices. Well-behaved AC performance in the form of clean transient response, symmetrical slew-rates and a high degree of forgiveness to supply decoupling are the hallmarks of this amplifier. AC gain and phase response are quite independent of temperature or supply voltage. Figure 1 shows the OP-42's small-signal response. Even with 75pF loads, there is minimal ringing in the output waveform. Large-signal response is shown in Figure 2. This figure clearly shows the OP-42's exceptionally close matching between positive and negative slew-rates. Slew-rate symmetry decreases the DC offset a system encounters when processing high-frequency signals, and thus reduces the DC current necessary for load driving.

FIGURE 1: Small-Signal Transient Response, $Z_1 = 2k\Omega||75pF$

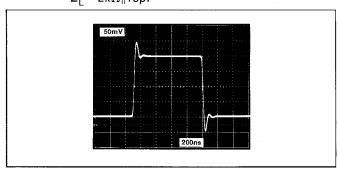
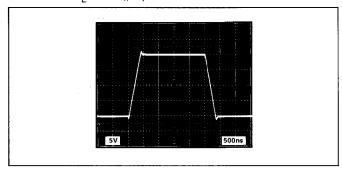


FIGURE 2: Large-Signal Transient Response, $Z_1 = 2k\Omega || 75pF$



As with most JFET-input amplifiers, the output of the OP-42 may undergo phase inversion if either input exceeds the specified input voltage range. Phase inversion will not damage the amplifier, nor will it cause an internal latch-up.

Supply decoupling should be used to overcome inductance and resistance associated with supply lines to the amplifier.

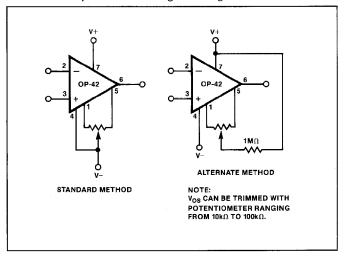
For most applications a 0.1μ F to 0.01μ F capacitor should be placed between each supply pin and ground.

The OP-42 displays excellent resistance to radiation. Radiation hardness data is available by contacting the factory.

OFFSET VOLTAGE ADJUSTMENT

Offset voltage is adjusted with a 10k Ω to 100k Ω potentiometer as shown in Figure 3. The potentiometer should be connected between pins 1 and 5 with its wiper connected to the V– supply. Nulling V_{OS} in this manner changes TCV_{OS} by no more than 5μ V/°C per millivolt of V_{OS} change. Alternately, V_{OS} may be nulled by attaching the potentiometer wiper through a 1M Ω resistor to the positive supply rail.

FIGURE 3: Input Offset Voltage Nulling



SETTLING-TIME

Guaranteed fast-settling is assured by sample-testing during production. The OP-42 is configured as a unity-gain follower in the test circuit of Figure 4. This test method has advantages over false-sum-node techniques in that the actual output of the amplifier is measured, instead of an error-voltage at the sum node. Common-mode settling effects are exercised in this circuit, in addition to the slew-rate and bandwidth effects measured by the false-sum-node method. A reasonably flat-top pulse is required as a stimulus.

The output waveform of the OP-42 being tested is clamped by Schottky diodes and buffered by the JFET source-follower. The signal is amplified by a factor of ten by the fast amplifier IC1, then Schottky-clamped before being output. The OP-41 provides overall offset nulling. Analysis of the waveform using a digitizing oscilloscope determines the op amp's settling-time.

FIGURE 4: Settling-Time Test Fixture

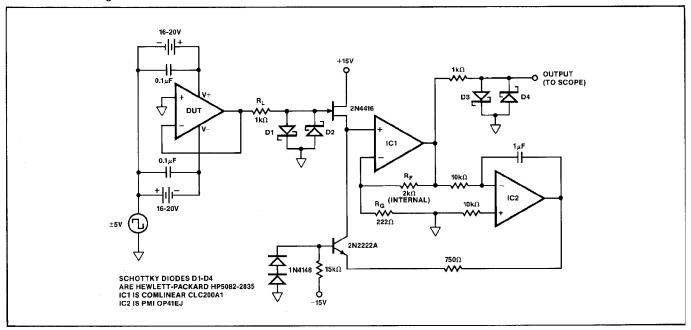
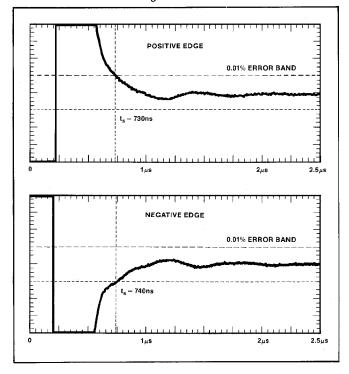


FIGURE 5: OP-42 Settling-Characteristics

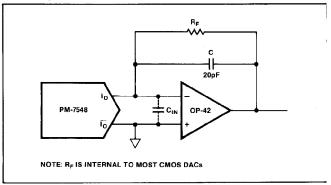


DAC OUTPUT AMPLIFIER

The OP-42 is an excellent choice for a DAC output amplifier, since its high speed and fast settling-time allow quick transitions between codes, even for full-scale changes in output level. The DAC output capacitance appears at the operational amplifier inputs, and must be compensated to ensure

optimal settling speed. Compensation is achieved with capacitor C in Figure 6. C must be adjusted to account for the DAC's output capacitance, the op amp's input capacitance, and any stray capacitance at the inputs. With a bipolar DAC, an additional shunt resistor may be used to optimize response. This technique is described in PMI's application note AN-24.

FIGURE 6: DAC Output Amplifier Circuit



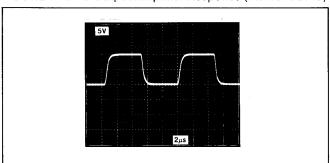
Highest speed is achieved using bipolar DACs such as PMI's DAC-08, DAC-10 or DAC-312. The output capacitances of these converters are up to an order of magnitude lower than their CMOS counterparts, resulting in substantially faster settling-times. The high output impedance of bipolar DACs allows the output amplifier to operate in a true current-to-voltage mode, with a noise gain of unity, thereby retaining the amplifier's full bandwidth. Offset voltage has minimal effect on linearity with bipolar converters.

CMOS digital-to-analog converters have higher output capacitances and lower output resistances than bipolar DACs.

This results in slower settling-times, higher sensitivity to offset voltages and a reduction in the output amplifier's bandwidth. These trade-offs must be balanced against the CMOS DAC's advantages in terms of interfacing capability, power dissipation, accuracy levels and cost. Using the internal feedback resistor which is present on most CMOS converters, the gain applied to offset voltage varies between 4/3 and 2, depending upon output code. Contributions to linearity error will be as much as 2/3V_{OS}. In a 10-volt 12-bit system, this may add up to an additional 1/5LSB DNL with the OP-42E. Amplifier bandwidth is reduced by the same gain factor applied to offset voltage, however the OP-42's 10MHz gain-bandwidth product results in no reduction of the CMOS converter's multiplying bandwidth.

Individual DAC data-sheets should be consulted for more complete descriptions of the converters and their circuit applications.

FIGURE 7: DAC Output Amplifier Response (PM-7545 DAC)



DRIVING A HIGH-SPEED ADC

The OP-42's open-loop output resistance is approximately 50Ω . When feedback is applied around the amplifier, output resistance decreases in proportion to open-loop gain divided by closed-loop gain (A_{VOL}/A_{VCL}). Output impedance increases as open-loop gain rolls-off with frequency. High-speed analog-to-digital converters require low source impedances at high frequency. Output impedance at 1MHz is typically 5Ω for an OP-42 operating at unity-gain. If lower output impedances are required, an output buffer may be placed at the output of the OP-42.

HIGH-CURRENT OUTPUT BUFFER

The circuit in Figure 8 shows a high-current output stage for the OP-42. Output current is limited by R1 and R2. For good tracking between the output transistors Q1, Q2 and their biasing diodes D1 and D2, thermal contact must be maintained between the transistor and its associated diode. If good thermal contact is not maintained, R1 and R2 must be increased to $5-6\Omega$ in order to prevent thermal runaway. Using 5Ω resistors, the circuit easily drives a 75Ω load (Figure 9). Output resistance is decreased and heavier loads may be driven by decreasing R1 and R2.

Base current and biasing for Q1 and Q2 are provided by two current sources, the MAT-02 and the JFET. The $2k\Omega$ potentiometer in the JFET current source should be trimmed for optimum transient performance. The case of the MAT-02 should be connected to V–, and decoupled to ground with a

 $0.1\mu F$ capacitor. Compensation for the OP-42's input capacitance is provided by C_C . The circuit may be operated at any gain, in the usual op amp configurations.

FIGURE 8: High-Current Output Buffer

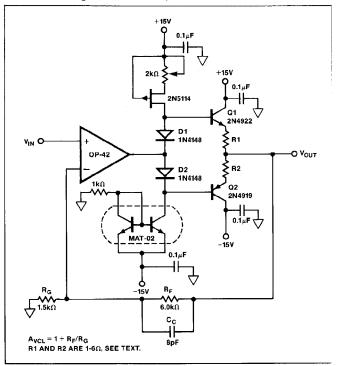
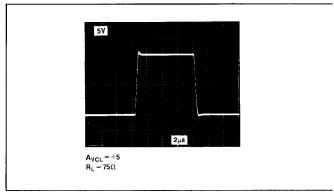


FIGURE 9: Output Buffer Large-Signal Response



DRIVING CAPACITIVE LOADS

Best performance will always be achieved by minimizing input and load capacitances around any high-speed amplifier. However, the OP-42 is guaranteed capable of driving a 100pF capacitive load over its full operating temperature range while operating at any gain including unity. Typically, an OP-42 will drive more than 250pF at any temperature. Supply decoupling does affect capacitive load driving ability. Extra care should be given to ensure good decoupling when driving capacitive loads, and a larger decoupling capacitor between $1\mu F$ and $10\mu F$ should be placed in parallel with the usual decoupling capacitor on each supply.

Large capacitive loads may be driven utilizing the circuit shown in Figure 10. R1 and C1 introduce a small amount of feedforward compensation around the amplifier to counteract the phase lag induced by the output impedance and load capacitance. At DC and low frequencies, R1 is contained within the feedback loop. At higher frequencies, feedforward compensation becomes increasingly dominant, and R1's effect on output impedance will become more noticeable.

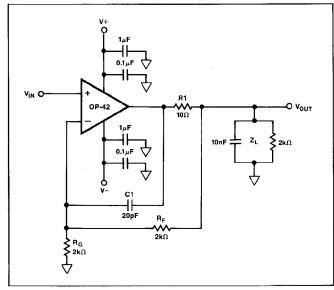
When driving very large capacitances, slew-rate will be limited by the short-circuit current limit. Although the unloaded slew-rate is insensitive to variations in temperature, the output current limit has a negative temperature coefficient, and is asymmetrical with regards to sourcing and sinking current. Therefore, slew-rate into excessive capacities will decrease with increasing temperature, and will lose symmetry.

COMPUTER SIMULATIONS

Many electronic design and analysis programs include models for op amps which calculate AC performance from the location of poles and zeros. As an aid to designers utilizing such a program, major poles and zeros of the OP-42 are listed below. Their location will vary slightly between production lots. Typically, they will be within $\pm 15\%$ of the frequency listed. Use of this data will enable the designer to evaluate gross circuit performance quickly, but should not supplant rigorous characterization of a breadboarded circuit.

POLES ZEROS 20Hz 1MHz 300kHz 3MHz

FIGURE 10: Compensation for Large Capacitive Loads

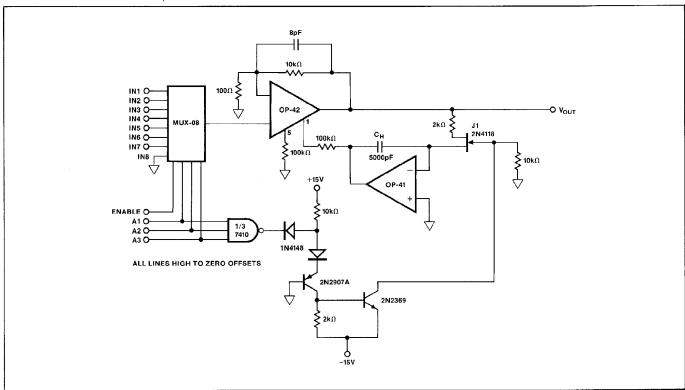


AUTOZEROING OFFSET VOLTAGE

Figure 11 describes a circuit for automatic offset voltage and drift correction. The OP-41 is used in a servo loop to force the OP-42 output equal to the OP-41's offset voltage. Thus, the OP-42's effective input offset is held below $10\mu V$ (1mV/A_{VCL} = 100) despite any temperature variations. This circuit will be most advantageous in high-gain applications.

Feedback is accomplished using the OP-42's null pins, leaving both inputs free for other purposes. In the application

FIGURE 11: OP-41 Servo Amplifier Provides Offset Correction



shown, the OP-42 has seven multiplexed inputs, while the eighth input provides a ground reference. Nulling is accomplished by addressing the grounded channel. This address should be held for at least $200\,\mu s$. After this time, the address may be changed to another channel. The MUX-08 ENABLE pin must be high during the entire nulling cycle. During this time, JFET switch J1 turns on, completing feedback around

the OP-41 servo amplifier. A charge is developed across C_H to compensate for the OP-42's offset voltage. When another channel is addressed, J1 turns off, and the correction charge is maintained across C_H by the OP-41. Droop is exceptionally low — only $1.3 \mu \text{V/s}$ at 25° C. A correction range of more than 4mV allows nulling of minor system offsets as well as the OP-42's offset voltage.